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Text: Nonlinear Control Systems — Analysis and Design, Wiley Author: Horacio J. Marquez

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Nonlinear Systems

- Nonlinear systems expression
 - Unforced, Autonomous, Example
- Equilibrium points
 - Examples first order, second order
- Phase-plane analysis
 - Vector field diagram
 - Examples
 - Limit cycle Van der Pol oscillator, MATLAB
 - Lorenz attractor MATLAB
- Exercises

• State x(t), Input u(t), Output y(t)

$$x(t) = \begin{bmatrix} x_1(t) \\ \vdots \\ x_n(t) \end{bmatrix}, \quad u(t) = \begin{bmatrix} u_1(t) \\ \vdots \\ u_p(t) \end{bmatrix}, \quad y(t) = \begin{bmatrix} y_1(t) \\ \vdots \\ y_m(t) \end{bmatrix}$$

• System: A set of first-order ordinary differential equations

$$\begin{aligned} \dot{x}(t) &= f(x(t), t, u(t)) \\ y(t) &= h(x(t), t, u(t)) \end{aligned}$$

i.e., (hereafter, (t) may be omitted)

$$\dot{x}_{1} = f_{1}(x_{1}, \dots, x_{n}, t, u_{1}, \dots, u_{p})$$

$$\vdots$$

$$\dot{x}_{n} = f_{n}(x_{1}, \dots, x_{n}, t, u_{1}, \dots, u_{p})$$

$$y_{1} = h_{1}(x_{1}, \dots, x_{n}, t, u_{1}, \dots, u_{p})$$

$$\vdots$$

$$y_{m} = h_{m}(x_{1}, \dots, x_{n}, t, u_{1}, \dots, u_{p})$$

Examples of Nonlinear Systems

• Unforced system: Input u is identically zero, i.e., u(t) = 0

$$\dot{x} = f(x, t, 0) = f(x, t)$$

• Autonomous system: f(x,t) is not a function of time

$$\dot{x} = f(x)$$

Autonomous systems are invariant to shifts in the time origin, i.e., changing the time variable from t to $\tau = t - \alpha$ does not change the right-hand side of the equation.

Example

- Input: *u*, Output: *x*
- Consider a system:

$$m\ddot{x} + d(x,\dot{x}) + k(x) = u$$

• $x_1 = x, x_2 = \dot{x}$

$$\dot{x}_1 = x_2$$

 $\dot{x}_2 = -\frac{d(x_1, x_2)}{m} - \frac{k(x_1)}{m} + \frac{1}{m}u$

i.e., the left-hand side should be first derivative of x, the right-hand side should not contain \dot{x} .

Example



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• **Definition**: A point $x = x_e$ is said to be an **equilibrium point** of the autonomous system

$$\dot{x} = f(x)$$

if it has the property that whenever the state of the system starts at x_e , it remains at x_e for all future time

$$x(t_0) = x_e \quad \Rightarrow \quad x(t) \equiv x_e, \ \forall t \ge t_0,$$

i.e.,

$$\dot{x} = 0.$$

• **Property**: The equilibrium points are the real roots of the equation $f(x_e) = 0$.

Example of Equilibrium Points

• Consider the following system where r is a parameter.

$$\dot{x} = -r + x^2$$

- 1. If r > 0, the system has two equilibrium points $x = \pm \sqrt{r}$.
- 2. If r = 0, both of the equilibrium points collapse, the equilibrium point is x = 0.
- 3. If r < 0, then the system has no equilibrium points.

$$\dot{x} = -r + x^2 \quad (r > 0)$$

When $\dot{x} > 0$, the trajectories move to the right, and vice versa. Thus $x_e = -\sqrt{r}$ is attractive, $x_e = \sqrt{r}$ is repelling.



First-Order Autonomous Systems II

 $\dot{x} = \cos x$

where $\dot{x} = 0$ are equilibrium points.



Second-Order Autonomous Systems

$$\dot{x}_1 = r - x_1^2$$
 (r = 0.5)
 $\dot{x}_2 = -x_2$



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Second-Order Autonomous Systems

$$\dot{x}_1 = r - x_1^2$$
 (r = 0.5)
 $\dot{x}_2 = -x_2$

where $(\sqrt{r}, 0)$ is stable, $(-\sqrt{r}, 0)$ is unstable.



```
[x, y] = meshgrid(-1.5:0.1:1.5, -1.5:0.1:1.5);
global r;
r = 0.5;
px=r-x.*x;
py=-y;
quiver(x,y,px,py,1.5);
[t,xx]=ode45(@func_bifur,[0 1.5],[0;1]);
plot(xx(:,1),xx(:,2),'go-');
function dx = func_bifur(t, x)
global r;
dx = [r-x(1)*x(1); -x(2)];
```

end

• Dynamical equation:







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• Consider the system

$$\dot{x}_1 = f_1(x_1, x_2)$$

 $\dot{x}_2 = f_2(x_1, x_2)$

- The solution of the differential equation with an initial condition $x_0 = [x_{10}, x_{20}]$ is called a **trajectory** from x_0 .
- The trajectory in x_1-x_2 plane is called **phase-plane**.
- f(x) in

$$\dot{x} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} f_1(x_1, x_2) \\ f_2(x_1, x_2) \end{bmatrix} = f(x)$$

is called a vector field.

Vector Field Diagram

- To each point x^* in the plane we can assign a vector with amplitude and direction of $f(x^*)$.
- For easy visualization we can represent f(x) as a vector based at x, i.e., we assign to x the directed line segment from x to x + f(x).
- Repeating this operation at every point in the plane, we obtain a **vector field diagram**.

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$$\dot{x}_1 = x_2$$

 $\dot{x}_2 = -x_1^2 - x_2$



Vector Field Diagram of Pendulum







 $(\pi/4,0), (\pi/2,0), (3\pi/4,0), (-\pi-\epsilon, 0.04), (\pi+\epsilon, -0.04)$

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Limit Cycles

- Oscillations: Characteristics of higher-order nonlinear systems
- A system oscillates when it has a **nontrivial** periodic solution (not the one found in LTI imaginary case).

$$\exists t_0 > 0, \quad \forall t \ge t_0, \quad x(t+T) = x(t)$$

• Stable, self-excited oscillations: limit cycles.

• Consider the following system:

$$\ddot{y} - \mu(1 - y^2)\dot{y} + y = 0$$
 $\mu > 0$

• Define $x_1 = y$, and $x_2 = \dot{y}$

$$\dot{x}_1 = x_2$$

 $\dot{x}_2 = -x_1 + \mu(1 - x_1^2)x_2$

• Note that if $\mu = 0$, the resulting system is

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

which is LTI and has circular trajectories.

Example of Limit Cycles (Van der Pol)









```
[x, y] = meshgrid(-4:0.4:4, -4:0.4:4);
global mu; mu=1;
px=y;
py=-x+mu*(1-x.*x).*y;
quiver(x,y,px,py,3);
[t,xx]=ode45(@vdp,[0 8],[-1;3]);
plot(xx(:,1),xx(:,2),'go-');
 _ _ _ _ _
function dx=vdp(t,x)
global mu;
dx = [x(2); -x(1) + mu * (1 - x(1) * x(1)) * x(2)];
end
```

Van der Pol Limit Cycle

- There is only one isolated orbit (Limit Cycle).
- All trajectories converge to this trajectory as $t \to \infty$, i.e., it is a **stable limit cycle**.

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Higer-Order Systems

• Chaos: Consider the following system of nonlinear equations (Ed Lorenz, 1963)

$$\dot{x} = \sigma(y - x)$$
$$\dot{y} = rx - y - xz$$
$$\dot{z} = xy - bz$$

where $\sigma, r, b > 0$.



Example of Limit Cycles (Lorenz attractor) 36

end

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Exercises

• For each of the following systems, (i) find the equilibrium points, (ii) plot the phase portrait, and (iii) classify each equilibrium point as stable or unstable.

(a)
$$\begin{cases} \dot{x}_1 = x_1 - x_1^3 + x_2 \\ \dot{x}_2 = -x_2 \end{cases}$$

(b)
$$\begin{cases} \dot{x}_1 = -x_2 + 2x_1(x_1^2 + x_2^2) \\ \dot{x}_2 = x_1 + 2x_2(x_1^2 + x_2^2) \end{cases}$$

(c)
$$\begin{cases} \dot{x}_1 = \cos x_2 \\ \dot{x}_2 = \sin x_1 \end{cases}$$